

Blind Linear and Nonlinear Mixture Identification Using Source Sparsity Assumptions

Matthieu Puigt

LISIC - ULCO

IUT de Saint-Omer Dunkerque

matthieu.puigt@lisic.univ-littoral.fr

<http://www-lisic.univ-littoral.fr/~puigt>

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Foundation for Research and Technology – Hellas (FORTH)



- FORTH
 - Public institution (1983) funded by projects
 - Largest Greek research centre, strong participation in EU projects
 - Seven research institutes in four different locations, one start-up incubator, one technology transfer organization, the Crete University press, the Observatory of Skinakas
- Institute of Computer Science (FORTH-ICS)
 - since 1983, internationally recognised, ERCIM member, 9 laboratories
- Signal Processing Lab (SPL)
 - 3 permanent researchers, 7 assistant researchers and 6 post-docs
 - Since 2005: 8 European Commission, 6 national, 3 industrial projects: more than 5.1 M€ in actual funding for FORTH. 6 active projects
 - Research interest: Statistical signal processing with applications in wireless sensor networks, image/video, audio and speech processing.

Outline of the talk

Part I

Basic introduction to blind source separation

Part II

Post-nonlinear sparse component analysis

Part III

Real-time source localization: a brief introduction

Let's talk about linear systems

All of you know how to solve this kind of systems:

$$\begin{cases} 2 \cdot s_1 + 3 \cdot s_2 & = & 5 \\ 3 \cdot s_1 - 2 \cdot s_2 & = & 1 \end{cases} \quad (1)$$

If we resp. define A , \underline{s} , and \underline{x} the matrix and the vectors:

$$A = \begin{bmatrix} 2 & 3 \\ 3 & -2 \end{bmatrix}, \underline{s} = [s_1, s_2]^T, \text{ and } \underline{x} = [5, 1]^T$$

Eq. (1) begins

$$\underline{x} = A \cdot \underline{s}$$

and the solution reads:

$$\underline{s} = A^{-1} \cdot \underline{x} = [1, 1]^T$$

Let's talk about linear systems

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$$\begin{cases} a_{11} \cdot s_1 + a_{12} \cdot s_2 & = & 5 \\ a_{21} \cdot s_1 + a_{22} \cdot s_2 & = & 1 \end{cases} \quad (1)$$

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How can we solve this kind of problem???

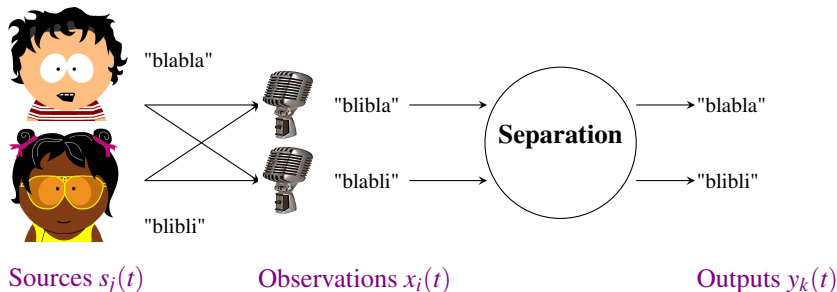
This problem is called **Blind Source Separation**.

Blind Source Separation problem

- N unknown sources s_j .
- One unknown operator \mathcal{A} .
- P observed signals x_i with the global relation

$$\underline{x} = \mathcal{A}(\underline{s}).$$

Goal: Estimating the vector \underline{s} , up to some indeterminacies.

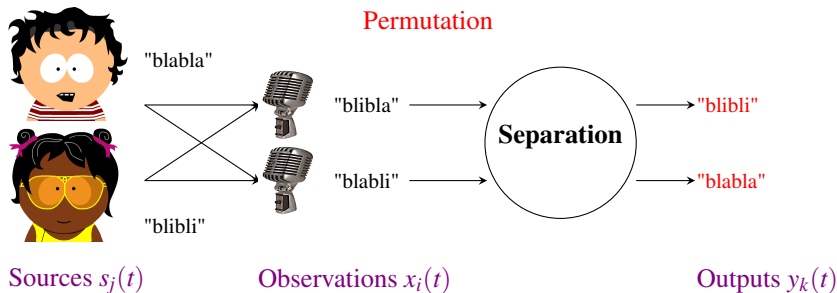


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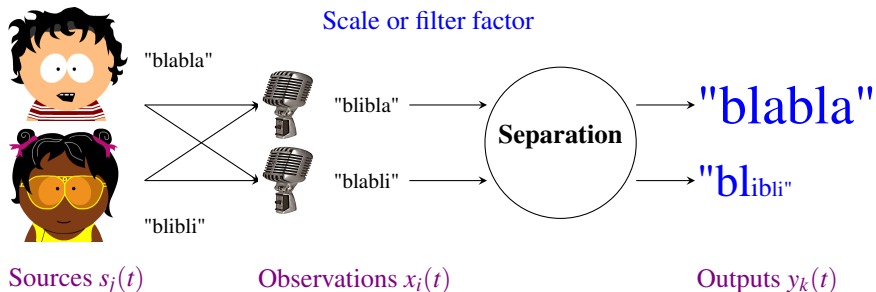


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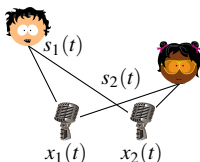
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Classes of mixtures

Most of the approaches process linear mixtures which are divided in three categories:

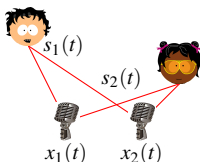
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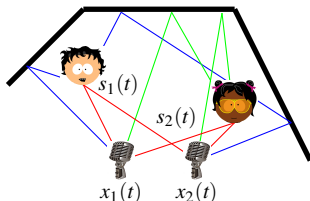


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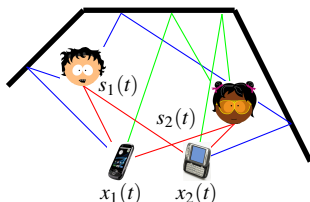


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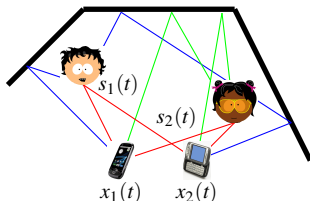
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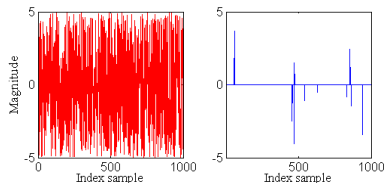
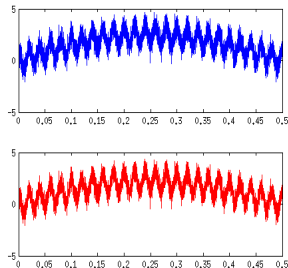
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- 5 Other classes of nonlinear mixtures (linear quadratic, polynomial, etc)

How to solve Blind Source Separation?

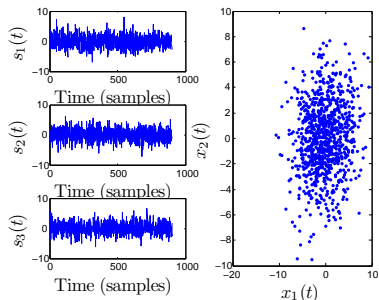
Three main families of methods:

- 1 **Independent Component Analysis (ICA):** Sources are statistically independent, stationary and at most one of them is Gaussian (in their basic versions).
- 2 **Sparse Component Analysis (SCA):** Sparse sources (i.e. most of the samples are null (or close to zero)). Purpose of this seminar
- 3 **Non-negative Matrix Factorization (NMF):** Both sources et mixtures are positive, with possibly sparsity constraints.



Sparse Component Analysis (1)

- Can process **underdetermined mixtures** (i.e. more sources than observations)

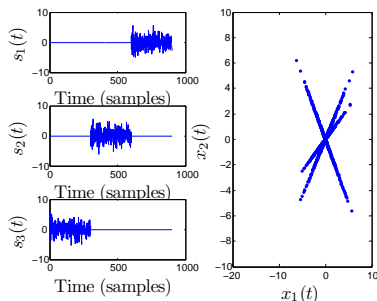


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 - If source signals are sparse in an analysis domain (e.g. time, Fourier, time-frequency, wavelet domains)
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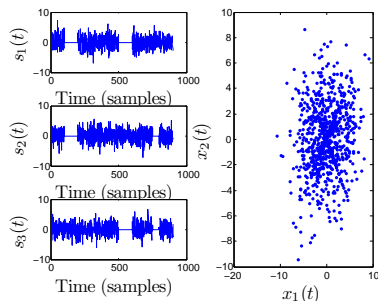
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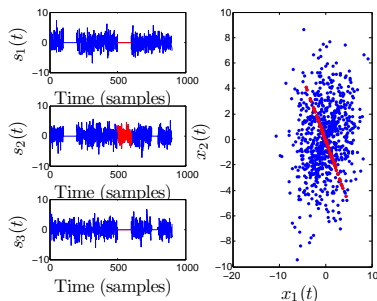


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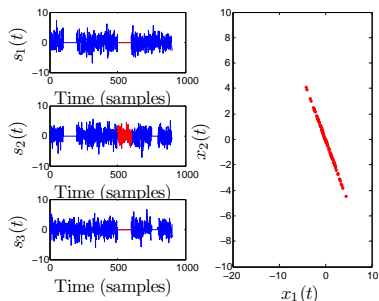


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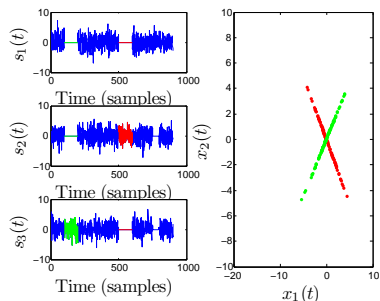


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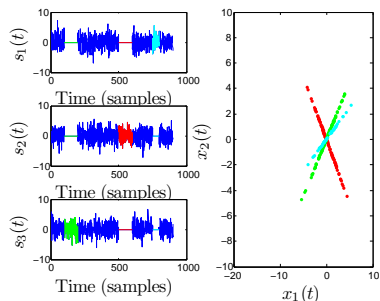


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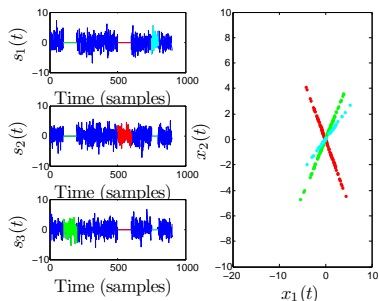


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- 2 Sources overlap, except if a few zones (to find) where only one of them is active (Deville *et al.*, 2001–2012)
- 3 $2 \leq Q < P$ sources are always active at each time (Cichocki *et al.*, 2004–2006)

Sparse component analysis (2)

Structure

Most of the SCA approaches follow the same structure:

- 1 Jointly sparsifying the observations $x_i(t)$
- 2 Estimating the mixing parameters:
 - e.g. by finding single-source zones using the ratio of observations (Deville *et al.*, 2001–2006), correlation (Deville *et al.*, 2004–2012), PCA (Arberet *et al.*, 2006–2010), the real and imaginary parts of the observations (Reju *et al.*, 2010), etc...
 - clustering estimates of the mixing parameters in the above zones (DEMIX, Selective K-means and K-medians, etc)
- 3 Estimating the sources (as an inverse problem)

Applications

As BSS is generic, SCA methods may be applied in numerous applications:

Audio domain

- Source cancelation (karaoke-like application)

[Observation 1](#)

[Observation 2](#)

[Output "without singer"](#)

- Separation for re-spatialization of the sound

Example (SiSEC 2008): [Observations](#) [Output 1](#) [Output 2](#) [Output 3](#)

- Audio enhancement (improving the perceptual sound of a speaker by removing the surrounding noise)

Example (SiSEC 2010): [Observations](#)

[Output 1](#)

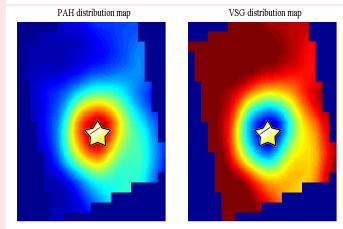
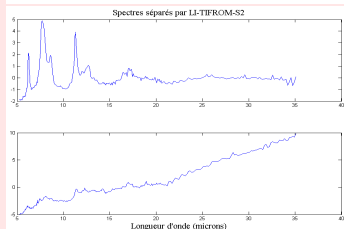
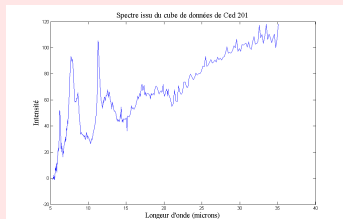
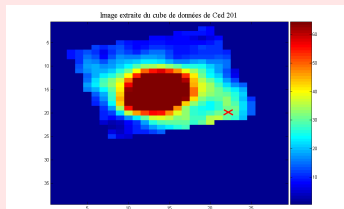
[Output 2](#)

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Astrophysical dust spectra from hyperspectral datacubes

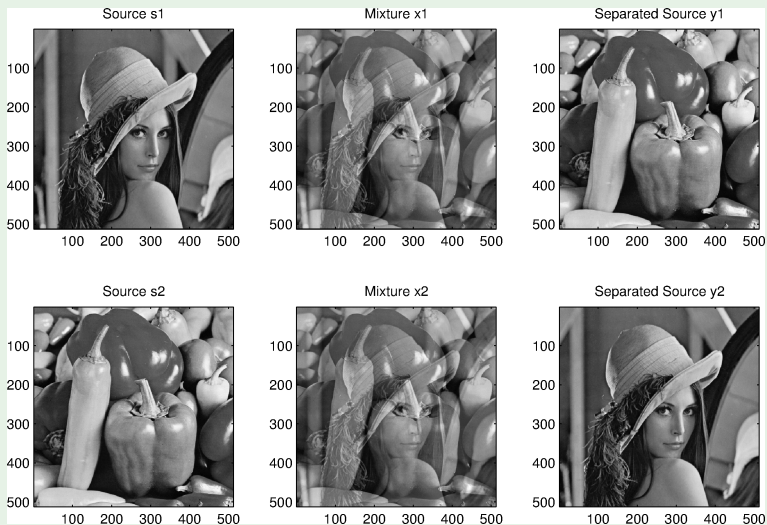
From Berné *et al.* (2007–2009) and Puigt *et al.* (2009).



Applications

As BSS is generic, SCA methods may be applied in numerous applications:

Images (Meganem *et al.*, 2010)



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Post-nonlinear sparse component analysis

Part III

Real-time source localization: a brief introduction

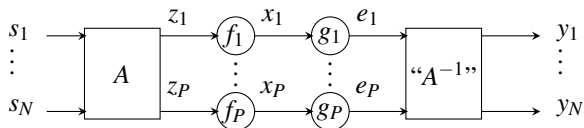
This work has been published in:

- M. Puigt, A. Griffin, and A. Mouchtaris, “Post-nonlinear speech mixture identification using single-source temporal zones & curve clustering”, in Proc. of EUSIPCO, pp. 1844-1848, 2011.
- M. Puigt, A. Griffin, and A. Mouchtaris, “Nonlinear blind mixture identification using local source sparsity and functional data clustering”, in Proc. of IEEE SAM, pp. 481-484, 2012.
- M. Puigt, A. Griffin, and A. Mouchtaris, “Post-nonlinear sparse component analysis using single-source zones and functional data clustering”, Preprint, <http://arxiv.org/abs/1204.1085>.

Post-nonlinear SCA: motivation

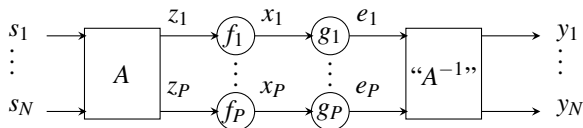
- Most of the people use their mobile device (e.g. smartphone) as a notebook: capturing images, videos, sounds and sharing them (social networks, streaming video websites, etc)
- But mobile devices are *mobile*:
 - Cheap and small microphone(s) & loudspeakers
 - ◇ provide **nonlinearities** in sound recording & restitution
 - Example: live recordings ([ex: found on Youtube](#))
 - Observed sound signals may be written as post-nonlinear (PNL) convolutive mixtures of source signals: $x_i(t) = f_i \left(\sum_{j=1}^N a_{ij}(t) * s_j(t) \right)$
- To tackle this problem, the literature provides:
 - 1 several ICA methods (limited to determined mixtures)
 - 2 a few SCA methods requiring the WDO assumption (Theis & Amari, 2004, van Vaerenbergh & Santamaría, 2006)
 - ◇ As we know that SCA outperforms ICA (Deville & Puigt, 2007), how to extend SCA methods to PNL convolutive mixtures?
 - ◇ In a first stage, how to extend LI-SCA methods to (instantaneous) PNL mixtures $x_i(t) = f_i \left(\sum_{j=1}^N a_{ij} \cdot s_j(t) \right)$?

Structure of the proposed method(s)



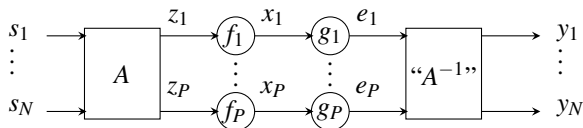
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 - 1 Estimate the inverse g_i of the NL functions f_i
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 - 1 Estimate the NL functions f_i
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 - 2 Find temporal single-source zones
 - 3 Estimate NL mappings
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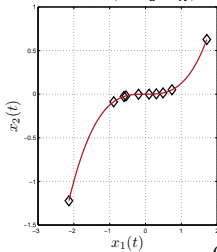
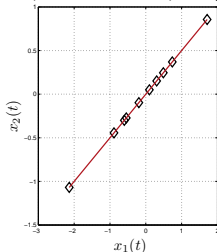
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What about the sparsifying transform?

Not a good idea (not sparser observations) \Rightarrow application: speech signals

Geometrical point of view

Let us imagine that, in one zone, only one source, say s_k , is active...



$$\forall i \in \{1, \dots, P\} \quad x_i(t) = f_i(a_{ik}s_k(t)) \Rightarrow s_k(t) = \frac{f_i^{-1}(x_i(t))}{a_{ik}}$$

and

$$x_i(t) = f_i \left(\frac{a_{ik}}{a_{1k}} f_1^{-1}(x_1(t)) \right) = \phi_{ik}(x_1(t))$$

Questions:

- How to find single-source zones?
- How to estimate ϕ_{ik} ?

Single-source confidence measures (1)

Mutual information (Puigt *et al.*, 2011)

- In linear mixtures, a source is isolated iff observations are proportional (correlation – Deville & Puigt, 2007)
- In PNL mixtures, need to measure the **nonlinear correlation** between observations.
- Done using mutual information (Dionisio *et al.*, 2004)

$$I(\underline{x}) = -\mathbb{E} \left\{ \log \frac{\prod_{i=1}^P \mathbb{P}_{x_i}(x_i)}{\mathbb{P}_{\underline{x}}(\underline{x})} \right\}$$

which may be normalised as $I_{\text{norm}}(\underline{x}) = \sqrt{1 - e^{-2I(\underline{x})}}$.

- ⇒ This implies that sources are mutually independent (✓ for speech—Puigt *et al.*, 2009)

Single-source confidence measures (2)

Manifold learning based measures (Puigt *et al.*, 2012a)

- Alternative to mutual information
- If the NL functions are smooth, then ϕ_{ik} are smooth and **locally linear**
- ⇒ We can *locally* apply linear measures (Manifold learning)
- Linear Tangent Space Approximation (LTSA—van der Maaten *et al.*, 2009) approximates the manifold around a value by its tangent in this point
- ① We consider each point t_i of a zone \mathcal{T} and we find its K -NN
- ② We apply a linear SSCM in this neighbourhood
 - correlation (Deville & Puigt, 2007): $C_{x_1, x_j}(t_i)$
 - ratio of eigenvalues (Arberet *et al.*, 2010): $R(t_i) = \frac{\lambda_1(t_i)}{\sum_{j=1}^K \lambda_j(t_i)}$
- ⇒ *Global* single-source confidence measure as the geometric mean of all the local SSCMs, respectively denoted $\mathcal{C}(\underline{x})$ and $\mathcal{R}(\underline{x})$

Single-source confidence measures (3)

Finally...

We look for zones \mathcal{T} such that $\text{SSCM}(\underline{x}) > 1 - \varepsilon_1$

Case of non-null unactive sources

- Problem may appear if inactive sources are **constant but non-zero** as $x_i(t) = f_i(a_{ik}s_k(t) + \alpha_i(T))$ where $\alpha_i(T) = \sum_{j \neq k} a_{ij}\bar{s}_j$
- Not a problem in Linear mixtures where observations may be centered...
- ➔ If all $f_i(0) = 0$ (not limiting assumption ✓), we discard all the estimated curves $\hat{\phi}_{ik}$ which do not satisfy $|\hat{\phi}_{ik}(0)| < \varepsilon_2$.

Functional data analysis and clustering (1)

- Different single-source zones may lead to scattered functions associated with the *same* source
- Interest to cluster all these zones in order to get an accurate estimate of the nonlinear mappings ϕ_{ik}
- Previously proposed methods also clustered data:
 - ① Theis and Amari: geometrical preprocessing sensitive to noise or **non-ideal** single source zones
 - ② Van Vaerenbergh and Santamaría: spectral clustering with curve distances limitations and which does not allow the curves to intersect (while all $f_i(0) = 0$) \Rightarrow **additional curve shape assumptions**
- We propose of taking advantage of the single-source zones: we estimate some parameters which adequately describe each scattered function and we cluster them.
- In particular, we test two families of methods:
 - ① We use a classical functional data clustering method
 - ② We propose a new approach well-suited to the considered problem.

Functional data analysis and clustering (2)

Filtering functional data clustering

Based on B-spline approximation (Abraham *et al.*, 2003): we approximate any nonlinear function with respect to a basis of polynomials

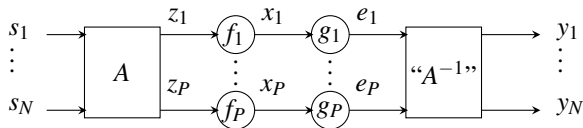
- 1 Select all the single-source zones and fix some knots locations
- 2 In each above zone, estimate $\hat{\phi}_{ik}$ using B-splines (all the B-spline coefficients have the same meaning and describe each curve shape)
- 3 Cluster the B-spline coefficients, in order to cluster the curves (K-medians)

Proposed method (Puigt *et al.*, 2012a)

Based on manifold learning around 0

- 1 For each zone, find the K -NN points around 0
- 2 Estimate the linear DOA in each zone with an approach of the literature (e.g. based on correlation or PCA)
- 3 Cluster these estimated DOAs (K-medians)

Final curve estimation and next steps of the separation approach



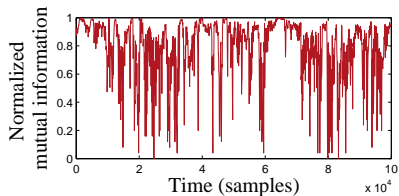
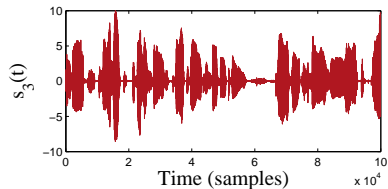
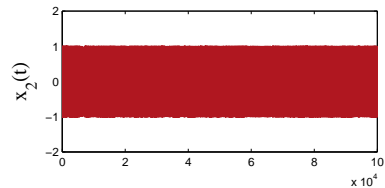
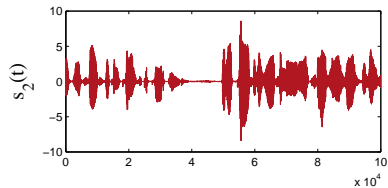
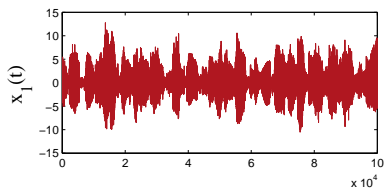
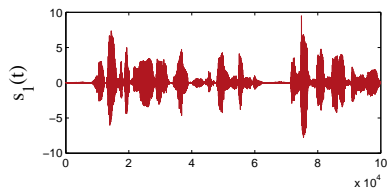
- We finally got separated functions that we e.g. can estimate thanks to B-splines, with much more knots and not-fixed locations.
- We can apply one approach of the literature to invert the nonlinearities (change the curves in lines).
- We then get a linear problem for which the mixing parameters can be estimated (slopes of the obtained lines).
- But we only focused on the nonlinearities estimation because *"it is of major importance for solving the BSS problem."*
- Let us see an example!

Example

- $N = 3$ sources (5 s, $F_s = 20$ kHz, silent parts) and $P = 2$ sensors
- PNL mixture: $A = \begin{bmatrix} 1 & 1 & 0.9 \\ -0.9 & 0.5 & 1 \end{bmatrix}$ and $\begin{cases} f_1(t) = \tanh(t) + t \\ f_2(t) = \tanh(10t) \end{cases}$
- Mutual information estimation with 100 samples per analysis zone, $\varepsilon_1 = 0.01$, $\varepsilon_2 = 0.1$

Example

Despite the strong NL, we find single-source zones

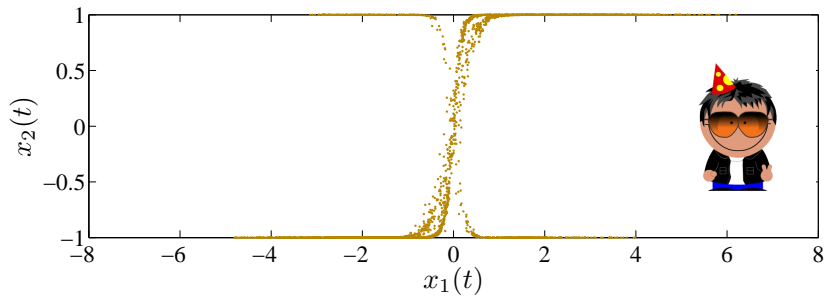
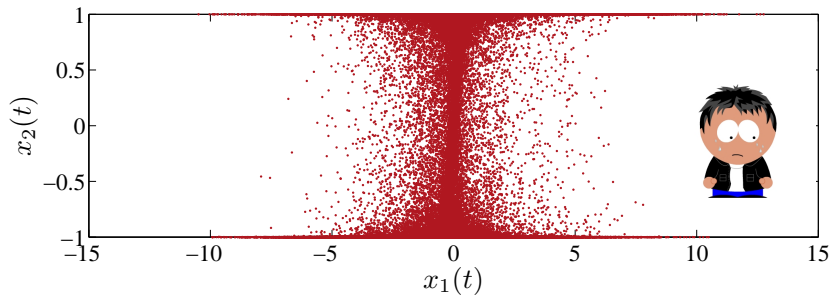


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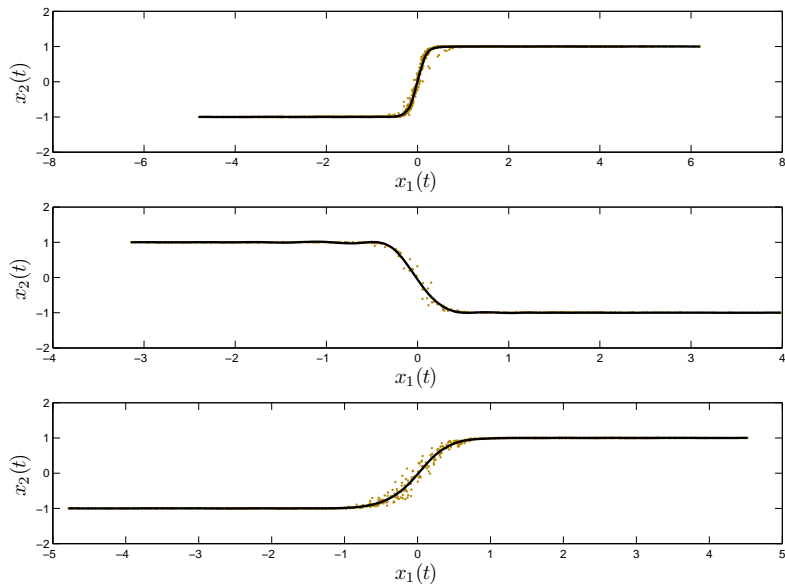
Example

On the influence of SSCMs



Example

Clustered data



Example

- $N = 3$ sources (5 s, $F_s = 20$ kHz, silent parts) and $P = 2$ sensors
- PNL mixture: $A = \begin{bmatrix} 1 & 1 & 0.9 \\ -0.9 & 0.5 & 1 \end{bmatrix}$ and $\begin{cases} f_1(t) = \tanh(t) + t \\ f_2(t) = \tanh(10t) \end{cases}$
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- Accurate classification and estimation, **MSE: 2.5e-4, 5.3e-5, and 2.1e-5**

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Detailed performance of the proposed approaches (Puigt *et al.*, 2012b)

We performed many tests (several NL functions, several mixing matrices while method's parameters vary) which show that:

- $I_{\text{norm}}(\underline{x})$ and $C(\underline{x})$ are better-suited than $\mathcal{R}(\underline{x})$
- Our manifold-learning clustering approaches are more flexible than B-spline functional data clustering one

Conclusion and future work

- A general framework for extending linear SCA to PNL mixtures (and even more general NL mixtures, see Puigt *et al.*, 2012a)
- Estimation of the non-linearities combines single-source zones with functional data clustering
- We proposed some Manifold-learning-based techniques for both tasks
- We also tested classical measures (mutual information and B-splines functional data clustering)
- Our results show that our approaches allow an accurate estimation of the nonlinearities
- We still have to invert them
 - May be done with an approach of the litterature
 - Or with a future proposed method...
- Our approach restricted to signals which are sparse in the time domain
 - ⇒ not well-suited to music
- Still need to investigate PNL convolutive mixtures

Part I

Basic introduction to blind source separation

Part II

Post-nonlinear sparse component analysis

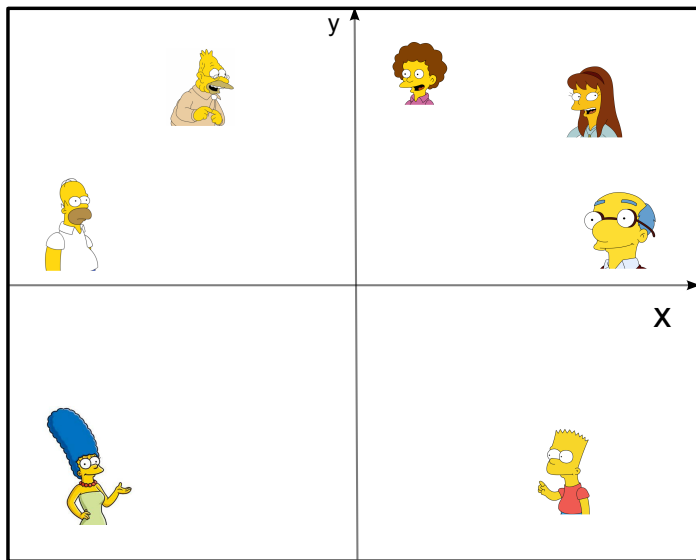
Part III

Real-time source localization: a brief introduction

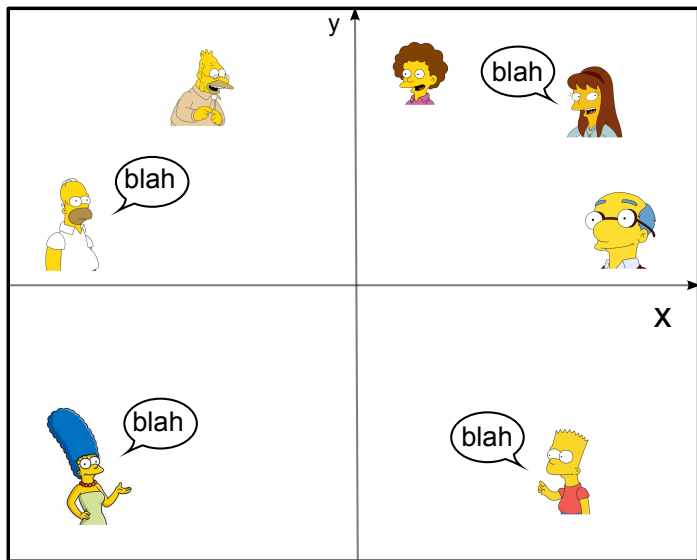
This work has been published in:

- D. Pavlidi, M. Puigt, A. Griffin, A. Mouchtaris, “Real-time multiple sound source localization using a circular microphone array based on single-source confidence measures”, in Proc. of IEEE ICASSP, pp. 2625-2628, 2012.
- D. Pavlidi, A. Griffin, M. Puigt, A. Mouchtaris, “Source counting in real-time sound source localization using a circular microphone array”, in Proc. of IEEE SAM, pp. 521-524, 2012.
- A. Griffin, D. Pavlidi, M. Puigt, A. Mouchtaris, “Real-time multiple speaker DOA estimation in a circular microphone array based on matching pursuit”, in Proc. of EUSIPCO, 2012.
- Journal paper just submitted to IEEE TASLP

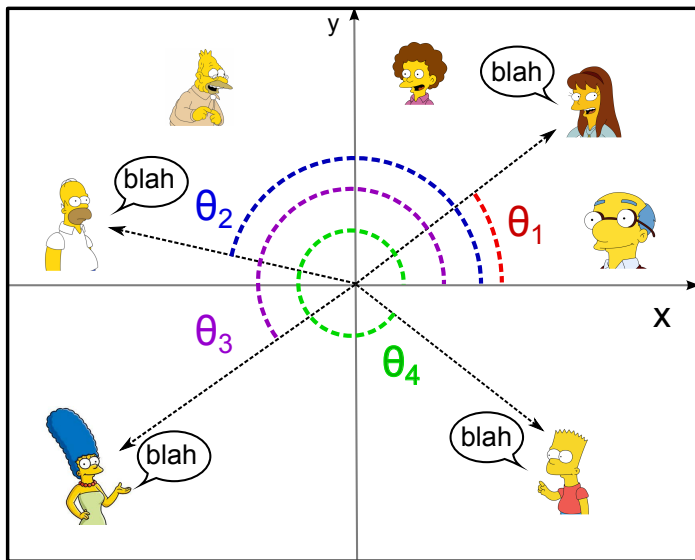
Motivation



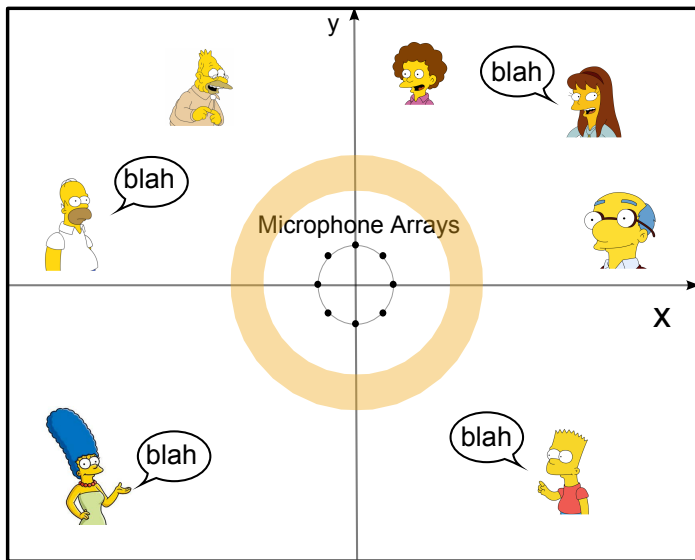
Motivation



Motivation



Motivation

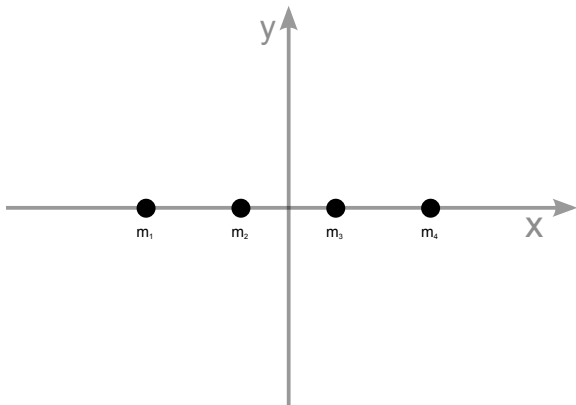


Microphone arrays

Geometry

Linear arrays

- Simple...

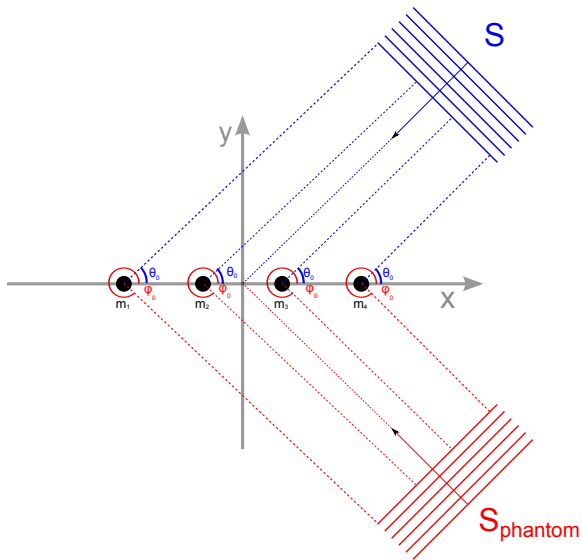


Microphone arrays

Geometry

Linear arrays

- Simple...
- But ambiguous!

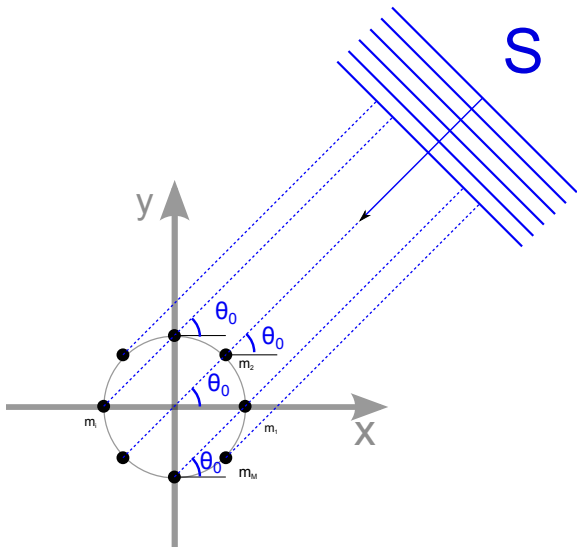


Microphone arrays

Geometry

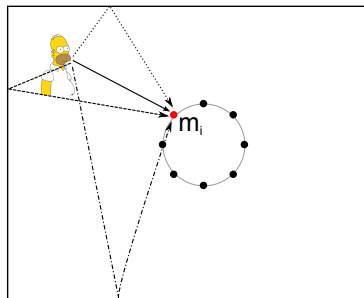
Circular arrays

Remove ambiguities!



Microphone arrays

Propagation Models



Reverberant Model

Time domain

$$x_i(t) = \sum_{g=1}^P h_{ig}(t) * s_g(t) + n_i(t)$$

Time-Frequency (TF) domain

$$X_i(t, \omega) = \sum_{g=1}^P H_{ig}(\omega) \cdot S_g(t, \omega) + N_i(t, \omega)$$

State of the Art

Single source localization

Based on Time difference of arrival

- the GCC family [Knapp 1976]

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Multiple source localization

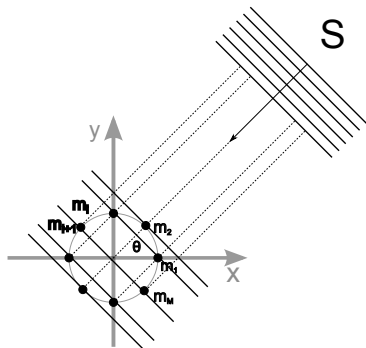
1 Based on statistics

- beamforming (2nd order statistics — e.g. multiple signal classification (MUSIC))[Argentieri 2007]
- Independent component analysis (2nd order or higher-order statistics) [Lombard 2008]

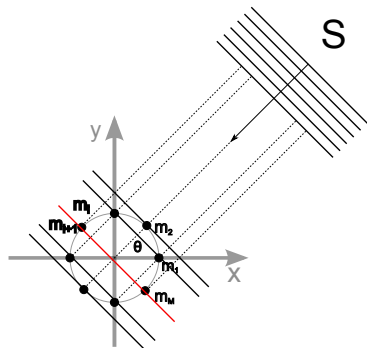
2 Using the sparsity paradigm

- Sparse component analysis (SCA) with W-disjoint orthogonality [Swartling 2006]
- Sparse component analysis with single-source confidence measure (framework of the present work)

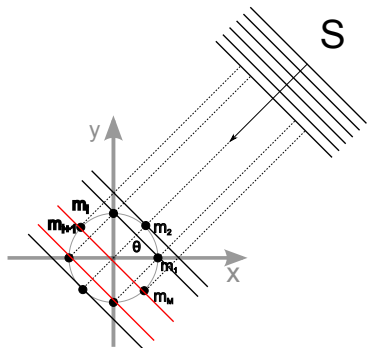
Single sound source localization using a circular microphone array [Karbasi 2007]



Single sound source localization using a circular microphone array [Karbasi 2007]

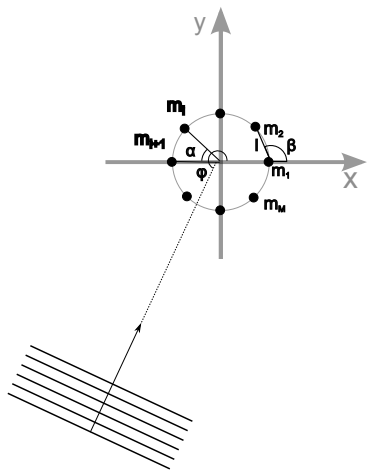


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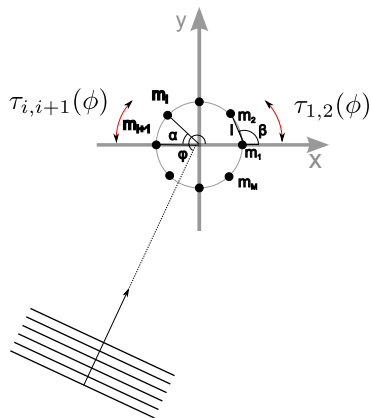
- $\tau_{i,i+1}(\theta)$
- $G_{i,i+1}^\theta(\omega) = \angle R_{i,i+1}(\omega) = e^{-j\omega\tau_{i,i+1}(\theta)}$

Single sound source localization using a circular microphone array [Karbasi 2007]



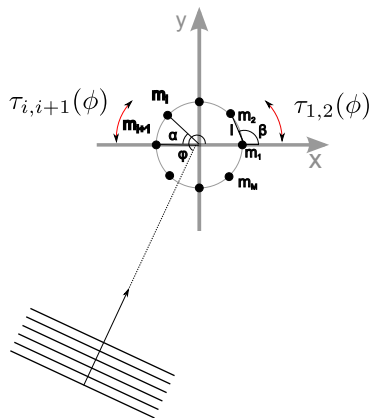
- $\tau_{i,i+1}(\phi) = \frac{l \sin(\beta - \phi + (i-1)\alpha)}{c}$
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- **Phase rotation factor:**
 $\text{PRF}_{m_i \rightarrow m_1}(\phi) = e^{-j\omega(\tau_{1,2}(\phi) - \tau_{i,i+1}(\phi))}$

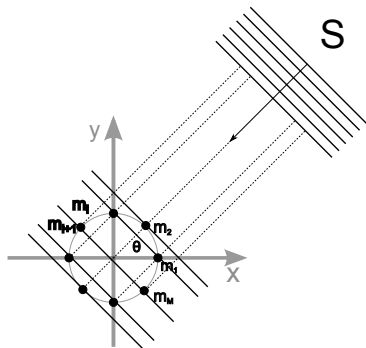
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$$\text{CICS}(\phi) \triangleq \sum_{i=1}^M \text{PRF}_{m_i \rightarrow m_1}(\phi) G_{i,i+1}^\theta(\omega)$$

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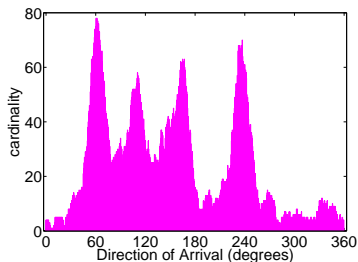
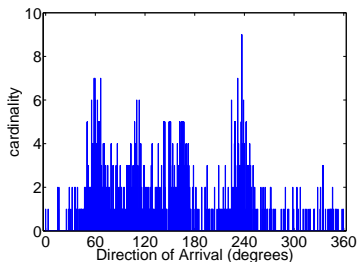
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$$\hat{\theta} = \arg \max_{0 \leq \phi < 2\pi} |\text{CICS}(\phi)|.$$

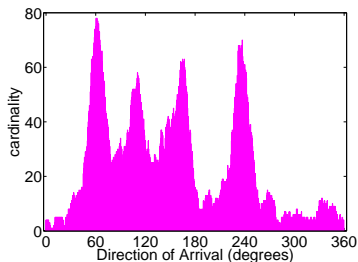
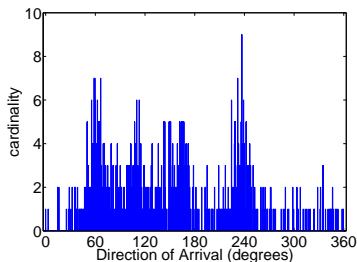
Multiple Sound Source localization method

- If we combine the linear SSCM (e.g. the correlation) with the above single-source localization approach, we provide a **real-time** multiple-source localization method (Pavlidis *et al.*, 2012a).
- Algorithm:
 - 1 We consider an history length of the signal (typically 1 s)
 - 2 We cut it in frames (typically 2048 points)
 - 3 We compute a FFT on each frame
 - 4 We find Constant-time single-source zones (Puigt & Deville, 2007)
 - 5 We estimate the associated DOA in each of these zones
 - 6 We derive a (smoothed) histogram of the above DOAs



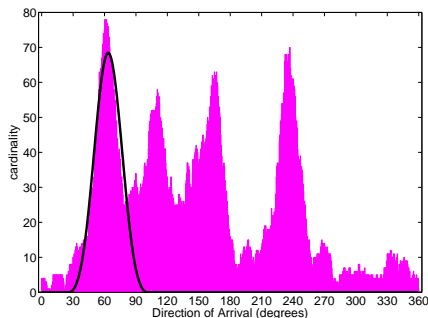
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 - 7 **We count the number of sources and we estimate their actual DOAs**



Source counting

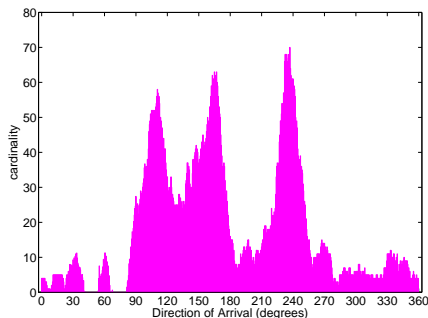
- We proposed 3 methods for source counting (Pavlidis *et al.*, 2012b)
- But only the most efficient one is here described
- Based on Matching Pursuit (model the histogram as linear combination of pulses)



- Define $\gamma = [\gamma_i], i = 1, \dots, P_{\text{MAX}}$.
- Correlate the source atom with the histogram
- Detect highest peak, set $i = 1$
- Calculate its contribution:
$$\delta_i = \sum_j \frac{y_{i,j} - y_{i+1,j}}{y_{1,j}}$$
- If $\delta_i \geq \gamma_i$, remove it, increment i .
- Continue iteratively until $\delta_i < \gamma_i$ (or $i = P_{\text{MAX}}$)

Source counting

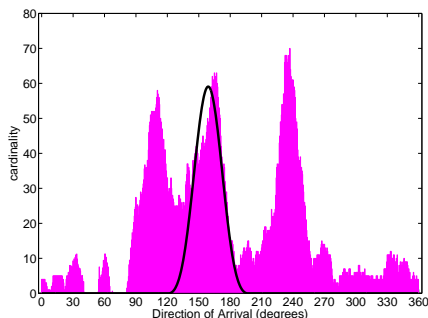
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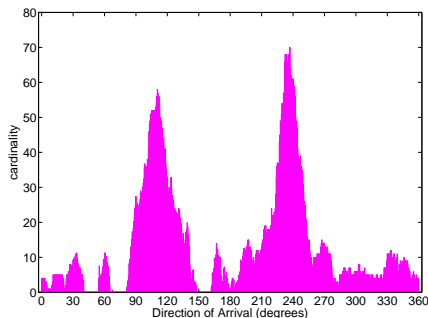
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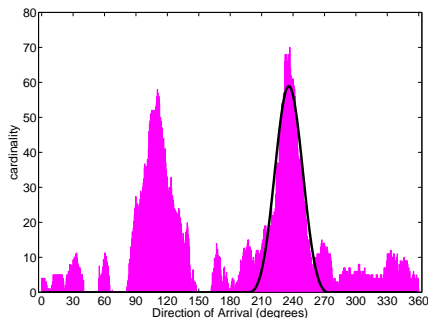
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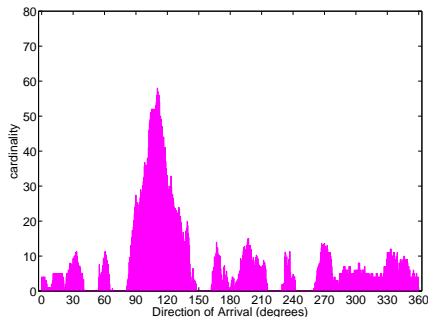
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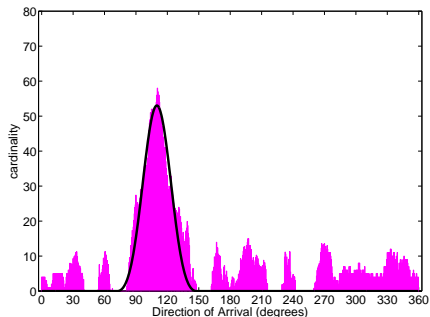
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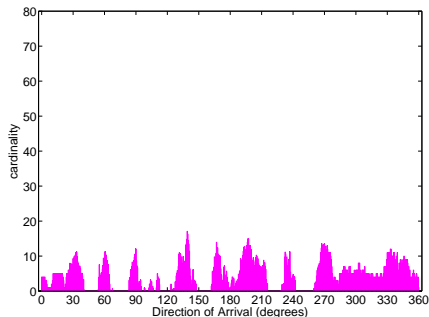
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- Correlate the source atom with the histogram
- Detect highest peak, set $i = 1$
- Calculate its contribution:
$$\delta_i = \sum_j \frac{y_{i,j} - y_{i+1,j}}{y_{1,j}}$$
- If $\delta_i \geq \gamma_i$, remove it, increment i .
- Continue iteratively until $\delta_i < \gamma_i$ (or $i = P_{MAX}$)

Source counting

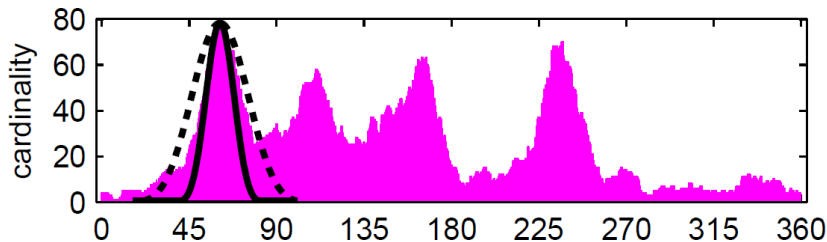
- We proposed 3 methods for source counting (Pavlidis *et al.*, 2012b)
- But only the most efficient one is here described
- Based on Matching Pursuit (model the histogram as linear combination of pulses)



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Final DOA estimation

- Two proposed methods: taking the indices of the highest value of the peaks or using Matching Pursuit (Griffin *et al.*, 2012)
- Key idea: The width of the above pulses is really important for getting accurate estimates (its shape may vary with the mixing conditions).
- Ideally, its optimal width should be estimated from the histogram (but time consuming)
- ◊ Combine two widths (was shown to be an acceptable trade-off)



- ◊ The whole method runs 55% real-time!

Performance (1)

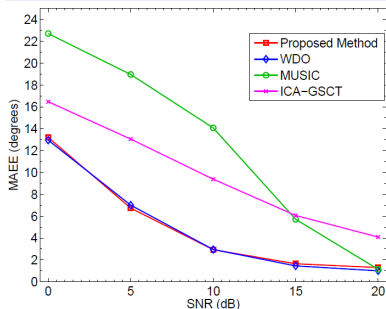
DOA estimation accuracy

- Simulations (6 sources, additive white noise, $T_{60} = 0.25$ s, room size: $4\text{m} \times 6\text{m} \times 3\text{m}$, 8 microphones) and comparison with Wideband Music, ICA-GSCT, and WDO-based method.
- Our approach is the less computational demanding (2.6E6 vs. between 3.9E6 & 35E6 operations)...

Performance (1)

DOA estimation accuracy

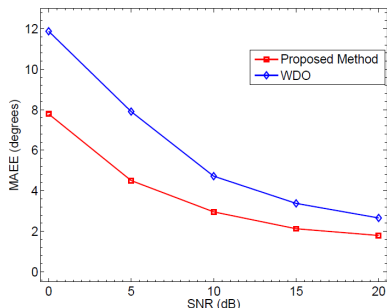
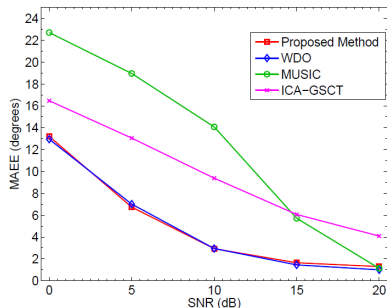
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- Our approach is the less computational demanding (2.6E6 vs. between 3.9E6 & 35E6 operations)...
- But it outperforms all the methods of the literature except the WDO one (perf. criterion: MAEE)
- It outperforms WDO in a 2-sources scenario



Performance (2)

Source counting accuracy

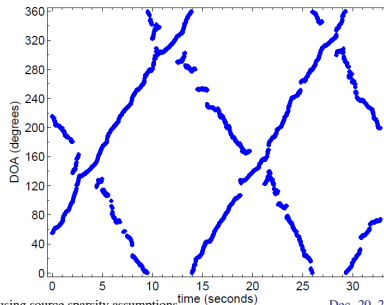
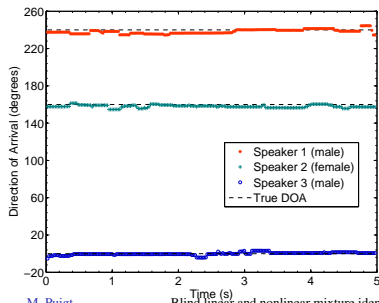
- Same simulated environment
- 4 intermittent sources
- Different history lengths (reponsiveness of the system)
- Different SNR conditions
- We measure a success rate of good source number estimation

History length (s)	SNR (dB)				
	0	5	10	15	20
0.25s	44.1%	60.2%	77.6%	85.0%	88.4%
0.5s	61.2%	81.7%	94.2%	96.0%	96.6%
1s	82.1%	99.2%	100%	100.0%	100.0%

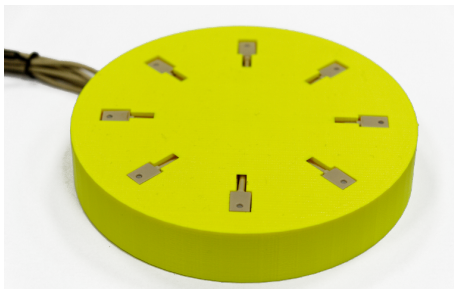
Performance (3)

Experiments in real environment

- Real speakers, static or moving around the array, who speak continuously
- Typical office room (same dimension as in simulations— $4\text{m} \times 6\text{m} \times 3\text{m}$) with A/C units ($\text{SNR} \approx 15\text{ dB}$)
- 8 omnidirectional Shure SM93 microphones, a TASCAM US2000 8-channel USB sound card, a Standard PC, Intel 3.00 GHz Core 2 CPU, 4 GB RAM, signal processing software in C++ and user interface in C#
- Good tracking of the sources



Conclusion



- Real-time multiple source counting and localization method (55% of the available time)
- Source sparsity (including WDO assumption) was shown to help getting good performance (even if the acoustic model is not that realistic!)
- SSCM helps to get more accurate DOAs and to be much faster!
- Left available time may be used e.g. for speaker diarization or separation
- Effects of hardware (cheap microphones)?

Thank you for your attention

Questions?